

METHODS AND ARRANGEMENTS FOR ENHANCING DOMINO LOGIC

FIELD OF INVENTION

5 **[0001]** The present invention is in the field of domino logic. More particularly, the present invention relates to methods and arrangements to enhance evaluation speed, reduce noise margin, and reduce leakage.

BACKGROUND

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15 **[0002]** As advances in semiconductor manufacturing technologies increase the number of transistors on a single die, computer architecture design focuses on increasing clock speeds. To reap the benefits of increased clock speeds, speed limitations associated with conventional logic must be overcome. In particular, evaluation time for synchronous logic must be reduced with increases in clock speed. Evaluation time is the time involved with generating a logical output in response to a logical input.

20 **[0003]** This continuous demand for performance improvement of integrated circuit (IC) chips has fueled the pervasive use of domino logic. When compared to conventional, complementary metal oxide semiconductor (CMOS) logic, domino logic offers high-speed operation, reduced die area, and reduced output capacitance.

25 **[0004]** Conventional domino logic includes a dynamic logic input circuit coupled with static gate circuits. A logic signal is received as an input on a first transition of a clock, and the signal is coupled to other circuitry and latched on the output on a next transition of the clock. More specifically, domino logic pre-charges a domino node, an input of the static circuitry, when a clock signal is low and couples the logic signal to the static circuitry via the dynamic logic input circuit when the clock signal is high. The dynamic logic input circuit often includes n-type metal oxide semiconductor (NMOS) transistors to pull down the voltage of the pre-charged,
30 domino node. Then, an output circuit changes states in response to the low charge on the

domino node to provide a stable output until the next clock cycle. Thus, the speed of domino logic is limited by the latencies involved with pre-charging the domino node, pulling down the pre-charged domino node voltage, and changing the state of the output circuit.

5 **[0005]** In addition, a keeper circuit is included in domino logic, compensating for leakage current, to maintain the pre-charge on the domino node. Leakage currents can cause an invalid evaluation when the domino node is discharged sufficiently by the leakage to change the output. The keeper circuit typically couples with the output and when the output indicates that the domino node is high, the keeper circuit is turned on to supply current to the domino node. The
10 keeper circuit, however, increases evaluation time because the dynamic logic input circuit must not only pull down the pre-charge on the domino node but also the current supplied by the keeper circuit.

[0006] Solutions for reducing the latencies typically reduce the physical channel length
15 of the circuit elements, the voltage supply for pre-charging the domino node, and the threshold voltages of selected transistors along a critical path within the domino circuit such as the threshold voltages of the transistors in the dynamic logic input circuit and the pull-up transistor in the output circuit. The short channel devices are able to change state over a smaller voltage range more quickly by driving more current. However, the reduced threshold voltages and
20 channel length of the transistors in the critical path result in higher subthreshold leakage currents. Thus, the keeper circuit must be sufficiently robust to handle the increased leakage currents, at the expense of increased evaluation time resulting from increased charge to be dissipated by the dynamic logic input circuit.

25 **[0007]** Further, the low threshold voltages along with faster signal edges increase problems associated with charge sharing and noise susceptibility. Charge sharing involves reduction of the voltage level at the domino node resulting from the capacitances associated with the domino node and the dynamic logic input circuit. Noise problems due to smaller geometry, often gauged by a measurement referred to as noise margin, result from noise coupling between,
30 e.g., the inputs of the dynamic logic input circuit and the domino node. For instance, a noise

introduced at the input of a low threshold voltage transistor in the dynamic logic input circuit may turn on the logic and drain current from the domino node, resulting in an invalid evaluation and output. Higher threshold voltage transistors can be implemented in the critical path through the domino circuit to compensate for higher noise, but such substitutions also increase evaluation
5 time.

SUMMARY OF THE INVENTION

[0008] The problems identified above are in large part addressed by methods and
10 arrangements for enhancing domino logic. One embodiment provides a circuit for domino logic wherein an output of the domino logic is based upon a charge on a domino node in response to a logic signal. The circuit generally includes a first keeper transistor having a gate coupled with the output; a second keeper transistor having a channel coupled in series with a channel of the first keeper transistor, the channels being between a high voltage source and the domino node;
15 and an input transistor having a gate coupled with the logic signal and a channel coupled between a gate of the second keeper transistor and the high voltage source.

[0009] Another embodiment provides an apparatus for enhancing domino logic. The
20 apparatus contemplates a logic input circuit to draw current from a domino node to generate a first logical output during a portion of a clock cycle; a clock circuit to couple the logic input circuit with a low voltage source during the portion of the clock cycle; an output circuit coupled to the domino node to output a second logical output when the domino node is above a threshold voltage; and a keeper circuit to pull up the domino node in response the output when the domino node is at a high voltage source and to stop pulling up the domino node before the output circuit
25 outputs the first logical output.

[0010] A further embodiment provides a method for enhancing domino logic. The
30 method can involve removing charge from a domino node via a first path in response to a logic signal; supplying charge from a high voltage source to a second node in response to the logic signal, the charge on the second node opening a second path to remove charge from the domino node; and generating an output based upon the removal of charge from the domino node.

BRIEF DESCRIPTION OF THE DRAWINGS

[0011] Other objects and advantages of the invention will become apparent upon reading
5 the following detailed description and upon reference to the accompanying drawings in which,
like references may indicate similar elements:

FIG 1 depicts an embodiment of a system including a clock, voltage sources, and logic;

FIG 2 depicts an embodiment of an apparatus including a clock circuit, a pre-charge
10 circuit, a logic input circuit, a keeper circuit, and an output circuit;

FIG 3 depicts an embodiment of an alternate clock circuit for the embodiment in FIG 2;

FIG 4 depicts an embodiment of alternate logic and mirror input circuits for the
embodiment of FIG 2;

FIGS 5-7 depict graphs of an embodiment displaying noise margin, charge sharing, and
15 speed enhancements for the embodiment in FIG 2; and

FIG 8 depicts an example of a flow chart for a method to enhance domino logic.

DETAILED DESCRIPTION OF EMBODIMENTS

20 [0012] The following is a detailed description of example embodiments of the invention
depicted in the accompanying drawings. The example embodiments are in such detail as to
clearly communicate the invention. However, the amount of detail offered is not intended to
limit the anticipated variations of embodiments, but on the contrary, the intention is to cover all
modifications, equivalents, and alternatives falling within the spirit and scope of the present
25 invention as defined by the appended claims. The detailed descriptions below are designed to
make such embodiments obvious to a person of ordinary skill in the art.

[0013] Generally speaking, methods and arrangements for enhancing domino logic are
contemplated. Embodiments include a keeper circuit to pull up a domino node in response an
30 output of an output circuit when the domino node is at a high voltage and to stop pulling up the
domino node before the output changes to a first logical output. In some of these embodiments,

a mirror input circuit couples with the keeper circuit to turn off, or block, the pull-up current before the voltage on the domino node falls below the threshold voltage of the output circuit. The mirror input circuit may also couple between a high voltage and an accelerator circuit to pull down the domino node while the logic input circuit is pulling down the domino node, to decrease evaluation time. In several of these embodiments, a clock circuit may substantially simultaneously couple the logic input circuit to a low voltage and the mirror input circuit to a high voltage via n-type, metal oxide semiconductor (NMOS) transistors. In other embodiments, particularly when the input logic signals are received after receiving a clock signal, a second clock signal may be generated to couple the mirror input circuit to a high voltage via a p-type, metal oxide semiconductor (PMOS) transistor, advantageously reducing the load on the clock signal.

[0014] Many embodiments use short channel transistors in the logic input circuit and the mirror input circuit to decrease evaluation time and, in one embodiment, the order of the transistors in the mirror input circuit is advantageously configured to block or partially block charge coupling introduced by the mirror input circuit when the worst cases of charge sharing is seen between the domino node and the logic input circuit.

[0015] Further embodiments include a pre-charge circuit that couples the keeper circuit with a low voltage source to pull down the voltage at an input of the keeper circuit while the domino node is high to maintain the pull up current and reduce or substantially suppress charge sharing associated with the mirror input circuit.

[0016] Turning now to the drawings, FIG 1 depicts an embodiment of a system 100 for enhancing domino logic. More specifically, system 100 may be a domino logic circuit including a clock 110; logic signals 115 and 160; a high voltage source 120; a low voltage source 130; logic 140, 150, and 170; and output 190. In many embodiments, system 100 may include other configurations of logic, voltage sources, and clocks.

[0017] Clock 110 may include circuitry to generate a clock signal to synchronize and propagate input logic signals 115 through the logic 140, 150, and 170. In particular, clock 110 may generate a signal that switches between a low voltage and a high voltage to define pre-charge cycles and switching cycles in logic 140, 150, and 170. For instance, when clock 110
5 outputs a low voltage signal, logic 140, 150, and 170 may pre-charge their respective domino nodes. When clock 110 transitions to a high signal, the pre-charge cycle may end and the logic signals 115 and 160 may be propagated through logic 140, 150, and 170 to generate outputs. In some embodiments, clock 110 may be configured to generate a clock signal having a frequency based upon and/or equivalent to another clock.

10 [0018] Logic signals 115 may include inputs to logic 140 and 150. Logic signals 160 may be outputs of logic 140 and 150 and inputs for domino logic 170. For example, during a first clock cycle, logic 140 and 150 may generate and latch logic signals 160. Then, during a second clock cycle, domino logic 170 may generate and latch output 190 based upon voltage
15 levels of logic signals 160.

[0019] High voltage source 120 may generate a high voltage for pre-charging domino nodes and for use as a high voltage logic signal. Low voltage source 130 may generate a low voltage for pulling down current from a domino node and for use as a low voltage logic signal.
20 For instance, during a first portion of a clock signal, domino logic 170 may pre-charge a domino node by coupling the domino node to high voltage source 120. Then, when logic signals 160 indicate a high voltage signal output, current may be drawn to low voltage source 130 from the domino node during the second portion of the clock signal and the high voltage source 120 may be latched to output 190.

25 [0020] Logic 140, 150, and 170 may represent two stages of logic within system 100. Each logic 140, 150, and 170 may include OR, NOR, AND, NAND, XOR, or similar logic operations. For example, domino logic 170 may perform a NAND logic operation with logic signals 160 to generate output 190.

[0021] Logic 140, 150 and 170 may each include circuitry depicted in domino logic 170 such as a clock circuit 171, a pre-charge circuit 172, a logic input circuit 173, a domino node 174, a keeper circuit 175, a mirror input circuit 178, an accelerator circuit 180, a pre-charge circuit 182, and an output circuit 180. Clock circuit 171 may receive a clock signal from clock 5 110 and turn on pre-charge circuit 172 when the clock signal is at a low voltage. Then, when the clock signal transitions to a high voltage, clock circuit 171 may turn off the pre-charge circuit 172, couple the logic input circuit 173 to the low voltage source 130, and couple a mirror input circuit 178 of keeper circuit 175 to high voltage source 120.

10 [0022] When turned on, pre-charge circuit 172 may supply current to domino node 174 from high voltage source 120 to raise the voltage on domino node 174 to or substantially to the voltage of high voltage source 120. When turned off, pre-charge circuit 172 may not supply current to or draw current from domino node 174.

15 [0023] Logic input circuit 173 and mirror input circuit 178 may receive logic signals 160 and determine whether to transition output 190 to a high voltage logic signal or leave the output as a low logic signal. For example, during the first portion of the clock signal when the clock signal is at a low voltage, pre-charge circuit 172 may be on to pull up domino node 174 to a high voltage. Then, during a second portion of the clock signal, logic input circuit 173 is coupled 20 with low voltage source 130. When logic input circuit 173 is designed to leave output 190 at a low voltage in response to logic signals 160, logic input circuit 173 draws no current or substantially no current from domino node 174. On the other hand, when logic input circuit 173 is designed to transition output 190 to a high voltage in response to logic signals 160, logic input circuit 173 draws current from domino node 174, pulling down the voltage of domino node 174 25 to or substantially to the voltage of low voltage source 130.

[0024] Keeper circuit 175 may supply current to domino node 174 to compensate for leakage currents, pulling up domino node 174 when domino node 174 is high and stopping or cutting off the current when logic input circuit 173 is pulling down domino node 174. In the

present embodiment, accelerator circuit 180 also pulls down the voltage on domino node 174 when logic input circuit 173 is pulling down the voltage on domino node 174.

[0025] Keeper circuit 175 may include pull-up 176. Pull-up 176 turns on when a low voltage is on node Y 177 and a low voltage is on output 190. Otherwise pull-up 176 turns off. When pull-up 176 is on, pull-up 176 couples domino node 174 to high voltage source 120, supplying current to domino node 174 to maintain domino node 174 at or substantially near the voltage of high voltage source 120. In other embodiments, pull-up 176 may turn on when a high voltage is on node Y 177 and a high voltage is on output 190.

[0026] Mirror input circuit 178 includes a folded stage of the input to domino logic 170, mirroring or substantially mirroring logic input circuit 173. In other words, mirror input circuit 178 couples with logic signals 160 and is designed to perform a complimentary logical operation for logic input circuit 173 since mirror input circuit 178 couples high voltage source 120 to node Y 177 of keeper circuit 175 rather than coupling low voltage source 130 to domino node 174. For instance, when logic input circuit 173 is designed to pull down domino node 174 in response to logic signals 160 via an AND operation, mirror input circuit 178 is designed to pull up node Y 177 of keeper circuit 175 in response to logic signals 160 via a NAND operation, pulling down domino node 174 by turning on accelerator circuit 180. Note that although logic input circuit 173 and mirror input circuit 178 may be comprise substantially the same configuration of transistors, logic input circuit 173 changes the charge on the domino node from high to low, to perform an AND operation, and mirror input circuit 178 changes the charge on node Y 177 from low to high, to perform a NAND operation.

[0027] When node Y 177 is pulled up by mirror input circuit 178, pull-up 176 is turned off and accelerator circuit 180 is turned on. Accelerator circuit 180 couples domino node 174 to low voltage source 130 while logic input circuit 173 is pulling down domino node 174 to advantageously increase evaluation speed. For example, when logic input circuit 173 begins to pull down domino node 174, mirror input circuit 178 pulls up node Y 177. Pull-up 176 turns off

and accelerator circuit 180 turns on, pulling additional charge from domino node 174 to low voltage source 130 in parallel with logic input circuit 173.

[0028] Pre-charge circuit 182 may pull down node Y 177 when domino node 174 is low to maintain a pull-up current via pull-up 176. On the other hand, when domino node 174 is pulled down below the threshold voltage for pre-charge circuit 182, pre-charge circuit 182 turns off, accelerating the pull-up of node Y 177 by mirror input circuit 178.

[0029] Output circuit 180 may couple output 190 with high voltage source 120 when the voltage or charge on domino node 174 is low and couple output 190 with low voltage source 130 when the voltage on domino node 174 is high. In other embodiments, output circuit 180 may couple output 190 with low voltage source 130 when the voltage or charge on domino node 174 is low and couple output 190 with high voltage source 120 when the voltage on domino node 174 is high.

[0030] Referring now to FIG 2, there is shown an embodiment of an apparatus 200 for enhancing domino logic. Apparatus 200 may include logic such as logic 170 in FIG 1. In particular, apparatus 200 may include a clock circuit 210, a pre-charge circuit 220, a logic input circuit 230, a domino node 240, a keeper circuit 250, a node Y 251, an accelerator circuit 257, a pre-charge circuit 259, a mirror input circuit 260, an output circuit 270, and an output 280. Clock circuit 210 may receive a clock signal 205 and coordinate synchronous operation of apparatus 200 based upon portions of a cycle of clock signal 205. Clock circuit 210 includes a coupling between clock signal 205 and pre-charge circuit 220, a low threshold, NMOS transistor 214 to couple logic input circuit 230 with a low voltage source, and a low threshold, NMOS transistor 216 to couple mirror input circuit 260 with a high voltage source. For example, when clock signal 205 is low, transistors 214 and 216 may be turned off, substantially isolating logic input circuit 230 from the low voltage source and mirror input circuit 260 from the high voltage source, respectively. On the other hand, when clock signal 205 is high, transistors 214 and 216 may be turned on, coupling logic input circuit 230 to the low voltage source and mirror input circuit 260 to the high voltage source, respectively. In other embodiments, clock circuit 210 may

include additional clock circuitry to generate a second clock signal as shown in and described in conjunction with FIG 3.

[0031] Pre-charge circuit 220 may pre-charge domino node 240 during a first portion of a cycle of clock signal 205. More specifically, pre-charge circuit 220 includes a high threshold, PMOS transistor 222. The gate of transistor 222 is coupled with clock signal 205 to turn on transistor 222 when clock signal 205 is low. The source of transistor 222 couples with a high voltage source and the drain of transistor 222 couples with domino node 240 to supply a current, or charge, to domino node 240, pulling up domino node 220 substantially to the voltage of the high voltage source.

[0032] Logic input circuit 230 may be designed to perform a logical operation like AND, NAND, OR, NOR or other operation and may include one or more transistors coupled with one or more logic signals. In the present embodiment, logic input circuit 230 is designed to perform an NAND operation with logic signals IN1-4. In particular, low threshold, NMOS transistors 232, 234, 236, and 238 are turned on when IN1-4 are high, respectively, and turned off when IN1-4 are low, respectively. For example, while clock signal 205 is low, transistor 222 pulls up domino node 240 to a high voltage. When clock signal transitions high, transistor 222 turns off and transistor 214 turns on, coupling transistors 232, 234, 236, and 238 to the low voltage source. If IN1-4 are high then current is drawn from domino node 240 to the low voltage source to drain the charge from domino node 240, pulling down domino node 240 substantially to the voltage of the low voltage source. On the other hand, when at least one of the logic signals IN1-4 is low, the corresponding transistor is turned off and substantially no current is drawn from domino node 240, except for current resulting from leakage, charge sharing, and noise coupling.

[0033] Keeper circuit 250 may pull up the domino node 240 in response to output 280 when domino node 240 is at a high voltage and stop pulling up domino node 240 before the output circuit 270 outputs the first logical output such as a high voltage. In the present embodiment, keeper circuit 250 also pulls down the voltage on domino node 240 while logic input circuit 230 is pulling down domino node 240. More specifically, keeper circuit 250

includes high threshold PMOS transistors **252** and **254** having channels coupled between the high voltage source and domino node **240** to supply current to domino node **240**, compensating for leakage and charge sharing. The source of transistor **252** couples with the high voltage source, the gate of transistor **252** couples with output **280** to turn transistor **252** on when the voltage on output **280** is low, and the drain of transistor **252** couples with the source of transistor **254**. The drain of transistor **254** couples with domino node **240** to supply current via transistor **252** to domino node **240** when both transistor **252** and transistor **254** are turned on.

[0034] Accelerator circuit **257** may include high threshold, NMOS transistor **256** having a channel coupled between domino node **240** and the low voltage source to pull down domino node **240** when logic input circuit **230** is pulling down domino node **240**, advantageously accelerating evaluation. In particular, the drain of transistor **256** couples with domino node **240** and the source of transistor **256** couples with the low voltage source. Node Y **251** couples with the gates of transistors **254** and **256** to turn off the pull up current and turn on the pull down current, or vice versa. For instance, when node Y **251** is low, transistor **256** is turned off and transistor **254** is turned on to allow a pull up current to increase the charge on domino node **240** to a high voltage. On the other hand, when node Y **251** is high, transistor **254** is turned off, blocking the pull up current and transistor **256** is turned on to draw current from domino node **240** to the low voltage source.

[0035] Many embodiments may include pre-charge circuit **259**. Pre-charge circuit **259** may include a high threshold, NMOS transistor **258** having a channel coupled between the node Y **251** and the low voltage source to pull down node Y **251** when domino node **240** is at a high voltage. The drain of transistor **258** couples with node Y **251**, the source of transistor **258** couples with the low voltage source, and the gate of transistor **258** couples with domino node **240**.

[0036] Mirror input circuit **260** may couple between the gates of transistors **254** and **256** and the high voltage source to turn off transistor **254** and to turn on transistor **256** when logic input circuit **230** draws current from domino node **240**. In the present embodiment, mirror input

circuit 260 includes a folded stage input of logic input circuit 230. More specifically, mirror input circuit 260 provides an AND operation on the same logic signals IN1-4 as logic input circuit 230, having transistors corresponding to each transistor of logic input circuit 230. For instance, when logic signals IN1-4 turn on transistors 232, 234, 236, and 238, the corresponding transistors of mirror input circuit 260, transistors 268, 266, 264, and 262, respectively, are also turned on. Thus, mirror input circuit 260 pulls up node Y 251 as logic input circuit 230 pulls down domino node 240.

[0037] Further, the order of the transistors in mirror input circuit 260 reduces charge sharing. In particular, transistor 232, the closest transistor to domino node 240, couples with logic signal IN1 in logic input circuit 230 but in mirror input circuit 260, transistor 268 couples with logic signal IN1 and is the farthest transistor away from node Y 251. As an illustration, assume that in a cycle of clock signal 205, domino node 240 is pre-charged to a high voltage node Y 251 is pre-charged low, logic signal IN4 is low, turning off transistors 238 and 262, and logic signals IN1-IN3 are high, turning on transistors 232, 234, 236, 268, 266, and 264. Charge sharing occurs between domino node 240 and junction capacitances 233, 235, and 237 associated with logic input circuit 230. However, as a result of the ordering of transistors in mirror input circuit 260, transistor 262 is turned off so charge sharing between junction capacitances 263, 265, and 267 and node Y 251 is suppressed and, in other circumstances, substantially attenuated. Further, in many embodiments, the width of the transistors 232, 234, 236, 238, 268, 266, 264, and 262 is smaller than conventional domino logic so the junction capacitances 233, 235, 237, 263, 265, and 267 are advantageously smaller.

[0038] Output circuit 270 couples with domino node 240 to output and latch a low voltage when the charge on domino node 240 is above a logic threshold voltage for output circuit 270, transistors 272 and 274, and to latch a high voltage when the charge on domino node 240 is below the logic threshold voltage for output circuit 270. Output circuit 270 may include low threshold voltage, PMOS transistor 272 and high threshold voltage, NMOS transistor 274, both having gates coupled with domino node 240. The source of transistor 272 couples with the high voltage source and the drain of transistor 272 couples with output 280 to supply a current to pull

up the charge of output **280** when transistor **272** is on. The drain of transistor **274** couples with output **280** and the source of transistor **274** couples with the low voltage source to pull output **280** down when transistor **274** is turned on. In other embodiments, different configurations of transistors may be used to generate and latch output **280**.

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[0039] Referring now to FIG 3, there is shown an embodiment of an alternate clock arrangement **300** for the embodiment in FIG 2. In particular, arrangement **300** includes the apparatus **200** of FIG 2 except clock B circuit **310** is added and high threshold, PMOS transistor **320** is substituted for mirror input circuit, clock transistor **216**. Clock signal **205**, logic input circuit **230**, mirror input circuit **260**, and logic input circuit, clock transistor **214** are shown to illustrate coupling between clock B circuit **310**, high threshold, PMOS transistor **320**, and the other elements of apparatus **200**.

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[0040] When clock signal **205** transitions to a high voltage, transistor **214** turns on and clock B circuit **310** generates an inverted, clock signal **315**. The inverted, clock signal **315** couples with the gate of high threshold, PMOS transistor **320** to turn on transistor **320**, coupling mirror input circuit **260** to the high voltage source. Generation of the inverted, clock signal **315** advantageously reduces the load on clock signal **205** since clock B can be used in multiple logics at the same time. In arrangement **300**, the input capacitance for clock signal **205** may be the effective capacitance of the gate capacitance for transistor **214** and clock B circuit **310**.

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[0041] Referring now to FIG 4, there is shown an embodiment of alternate logic input and mirror input circuits **430** and **460**, respectively, for apparatus **200** of FIG 2. In particular, arrangement **400** includes the apparatus **200** of FIG 2 except logic input circuit **230** is replaced with logic input circuit **430** between domino node **240** and logic input circuit, clock transistor **214** and mirror input circuit **260** is replaced with mirror input circuit **460** between node Y **461** and logic input circuit, clock transistor **216**.

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[0042] Logic input circuit **430** includes four transistors **432**, **434**, **436**, and **438**, providing four parallel paths for current between domino node **240** and the low voltage source when the

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transistors **432**, **434**, **436**, and **438** are turned on by logic signals **IN1-4**, respectively. For example, when logic input circuit, clock transistor **214** is turned on by a high voltage of clock signal **205**, logic signal **IN1** is high, and logic signals **IN2-4** are low, current may be drawn from domino node **240** through transistor **432** and transistor **214** to the low voltage source.

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[0043] Mirror input circuit **460** includes complimentary logic, four parallel transistors **462**, **464**, **466**, and **468** coupled with **IN1-4** respectively, to pull up node **Y 461** as logic input circuit **430** pulls down domino node **240**. For example, when transistor **432** pulls down domino node **240**, transistor **462** pulls up node **Y 461**.

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[0044] FIGS 5-7 depict graphs of an embodiment displaying noise margin, charge sharing, and speed enhancements. In particular, FIG 5 depicts the improvement to in recovery from charge sharing for the apparatus **200** in FIG 2 over conventional domino logic that uses high threshold transistors in a logic input circuit like logic input circuit **230** when logic signals, **IN1-3**, transition high and logic signal **IN4** is low. Input represents logic signals **IN1-3**. Domino node (CONV) describes the loss and recovery of charge on the domino node of the conventional domino logic input circuit. And domino node (NSDL) describes the loss and recovery of charge on domino node **240** of FIG 2. Note that the loss is less and thus, the recovery is faster in apparatus **200**. The loss may be less as a result of the lower junction capacitances, **233**, **235**, and **237** and the lower junction capacitances result from the smaller width of low threshold transistors **232**, **234**, **236**, and **238** in the in logic input circuit **230** as compared to the width of transistors in the corresponding logic input circuit for the conventional domino logic.

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[0045] FIG 6 depicts the improvement in noise response for the apparatus **200** in FIG 2 over conventional domino logic and dual voltage threshold (dual V_t) domino logic that uses low voltage threshold transistors in the logic input circuit like logic input circuit **230** and in the output circuit like output circuit **270** as a pull-up transistor. Various amplitudes of noise are introduced on the domino nodes of the conventional domino logic, dual V_t domino logic, and apparatus **200** with a 100 picosecond pulse. Domino node (CONV), domino node (DUAL V_t), and domino

node (NSDL) represent the affect of the noise on the domino nodes of the conventional domino logic, dual Vt domino logic, and apparatus 200, respectively. Note that even as the amplitude of the noise increases, the output node of apparatus 200, output (NSDL) remains more resistant to the noise than the conventional domino logic, output node (CONV), and the dual Vt logic, output node (DUAL Vt).

[0046] FIG 7 depicts the improvement in evaluation speed for the apparatus 200 in FIG 2 over conventional domino logic and dual voltage threshold domino logic described for FIGS 5 and 6. Input describes the rise of logic signals that cause logic input circuits to begin pulling down the respective domino nodes. Domino node (NSDL) represents the discharge of domino node 240 in response to the logic signals, which drops faster than the charge on the domino nodes of conventional and dual Vt domino logic, domino node (CONV) and domino node (DUAL Vt), respectively. Similarly, the charge on the output of apparatus 200, out(NSDL), rises faster than the charge on the outputs of conventional and dual Vt domino logic, out(CONV) and out(DUAL Vt), respectively.

[0047] Referring now to FIG 8, there is shown an example of a flow chart 800 for a method to enhance domino logic. In particular, flow chart 800 describes a method for implementing enhanced domino logic in a circuit like apparatus 200 of FIG 2, including a domino node coupled with a pre-charge circuit, a logic input circuit, an output circuit, and a keeper circuit, and a node Y coupled between a mirror input circuit and the keeper circuit to adjust the operation of the keeper circuit. Flow chart 800 begins with element 810, receiving a first clock signal transition. The first clock signal transition may include a low to high transition or a high to low transition and is received by a clock circuit configured to synchronize pre-charge and switching stages of progression wherein the pre-charge stage pre-charges the domino node and the switching stage generates and latches an output based upon the logic signals received as input. The first clock signal transition, in this embodiment, is the transition indicating progression from the pre-charge stage to the switching stage. In further embodiments, receiving the first clock signal transition may initiate generation of a second clock signal to turn on a logic

input circuit, clock transistor and/or a mirror input circuit, clock transistor, advantageously reducing input capacitance of and the load on the clock signal.

[0048] Upon receipt of the first clock transition, the logic input circuit and the mirror
5 input circuit are turned on (element 815). Turning on the logic input circuit may involve applying a voltage to the gate of a logic input circuit, clock transistor. The difference between the voltage applied and the voltage coupled with the source of the transistor may be greater than or equal to the threshold voltage for the transistor. The transistor is turned to facilitate a current
10 between the domino node and the low voltage source via the channel of the transistor when the logic input circuit permits current between the domino node and the low voltage source (element 820). Similarly, turning on the mirror input circuit may involve applying a voltage to the gate of a mirror input circuit, clock transistor to turn on that transistor (element 820).

[0049] Current is facilitated by the logic input circuit and the mirror input circuit when
15 the transistors of the logic input circuit, along a path between the domino node and the low voltage source, are turned on as a result of the logic signals received at the gates of the transistors of the logic input circuit. Similarly, current is facilitated by the mirror input circuit when the transistors of the mirror input circuit, along a path between the node Y and the high voltage source, are also turned on as a result of the logic signals applied to the gates of the transistors of
20 the mirror input circuit. Otherwise, the domino logic allows the present output to remain and waits for the next clock signal transition (element 855).

[0050] As current is drawn through the logic input circuit from the domino node to the low voltage source, pulling down the domino node (element 825), current is also drawn
25 substantially simultaneously through the mirror input circuit from the high voltage source to node Y, pulling up node Y (element 830). Node Y couples with the gate of a first keeper pull-up transistor and an accelerator transistor. Once the voltage on node Y reaches a threshold voltage for the first keeper pull-up transistor, the pull-up current for the domino node is blocked (element 835). Then, when the voltage on node Y reaches the threshold voltage of the accelerator circuit,

the accelerator transistor is turned on and the current is drawn from the domino node to the low voltage source to accelerate evaluation (element **840**).

[0051] Evaluation is dependent upon the time to reduce the voltage, or charge, on the domino node below the threshold voltage of a transistor in the output circuit. The output circuit may include a pull-up transistor having a gate coupled with the domino node and a channel coupled between the high voltage source and the output. The output circuit may also include a pull-down transistor having a gate coupled with the domino node and a channel coupled between the low voltage source and the domino node. As the pre-charge on the domino node is reduced, the voltage drops below the threshold voltage of the pull-down transistor for the output circuit and the pull down transistor turns off (element **845**). Then, when the voltage of the domino node reaches the threshold voltage of the pull-up transistor, the pull-up transistor turns on to pull up and latch the output (element **850**). Note that the threshold voltage of the pull-up transistor may be substantially equivalent to the threshold voltage of the pull-down transistor but offset slightly to avoid leakage directly from the high voltage source to the low voltage source through the channels of the pull-up and pull-down transistors in the output circuit. Further, the output may couple with the gate of a second keeper, pull-up transistor. Switching the output from low to high may turn off the second keeper, pull-up transistor, blocking pull-up current from the high voltage source, through the keeper pull-up circuit, to the domino node.

[0052] Once the output is latched, the domino logic may receive the second clock transition such as a high to low or low to high transition (element **855**). Typically, the second clock transition is opposite to the first clock transition, changing the clock signal to the opposite voltage level. However, other transitions are also contemplated.

[0053] The second clock transition indicates a change from the switching stage to the pre-charge stage. In response to the transition, the logic input circuit and the mirror input circuit are substantially blocked from conducting current (element **860**) although current may result from charge sharing, leakage, and noise coupling. Also, the pre-charge circuit is turned on to

conduct current from the high voltage source to the domino node, pre-charging the domino node to a high voltage (element 865).

[0054] Once the voltage reaches the threshold voltage to turn on a node Y, pre-charge transistor having a channel coupled between the low voltage source and node Y, node Y is pulled down to a voltage substantially equivalent to the voltage of the low voltage source (element 870). Node Y, being coupled to the gates of the first keeper pull-up transistor and the accelerator transistor, turns on the first keeper pull-up transistor, and turns off the accelerator transistor, blocking a path between the domino node and the low voltage source.

[0055] After the first keeper pull-up transistor is turned on and the accelerator transistor is turned off, the voltage of the domino node may rise sufficiently to turn off the pull-up transistor of the output circuit and turn on the pull-down transistor of the output circuit, transitioning the output from high to low (element 875).

[0056] When the output transitions from high to low, the second keeper pull-up transistor is turned on, supplying current from the high voltage source to the domino node via the channels of the first and second keeper pull-up transistors (element 880). The pull-up current supplied to the domino node may be designed to compensate for losses of charge resulting from, e.g., charge sharing and leakage.

[0057] It will be apparent to those skilled in the art having the benefit of this disclosure that the present invention contemplates methods and arrangements for enhancing domino logic by adding circuitry to make use of short channel devices and/or circuitry to accelerate evaluation. It is understood that the form of the invention shown and described in the detailed description and the drawings are to be taken merely as examples. It is intended that the following claims be interpreted broadly to embrace all the variations of the example embodiments disclosed.